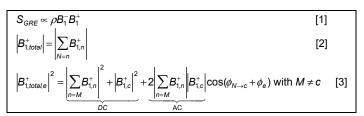
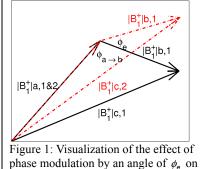
Fast phase-modulated B1+ mapping in the low flip-angle regime

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Introduction and purpose: In high-field MRI phased transmit arrays are used to mitigate RF issues, e.g. optimizing the inhomogeneous B₁⁺ field using spatially tailored RF pulses played out by such an array. In order to design the RF pulses that produce the desired excitation field, the B₁⁺ field per coil must be mapped. Ideally, these B1 mapping techniques are fast and have a low SAR deposition. However, most of the contemporary B1 mapping techniques require large flip angles [1,2] and are therefore accompanied by relatively high SAR values. As a consequence, lengthy acquisition times are needed to stay within SAR limits. In this abstract we describe a fast method to measure the B₁⁺ field per coil in the low flipangle regime, using phase modulated 3D low flip angle (PMLF) GRE images.

Theory: In GRE images for low flip angles, the signal amplitude is directly proportional to the product of the proton density, the receive sensitivity (B_1^-) and the transmit sensitivity (B_1^+) [equation 1], where the B_1^+ field is the superposition of the individual coil fields [equation 2]. The relation in equation 1 invokes that in order to disentangle B₁⁺ from the other two factors, a nonlinear manipulation of the transmit sensitivity is needed. Here we propose to achieve this by electronically phase modulating one of the coil input signals with angle $\phi_{\rm e}$. The resulting ${\rm B_1}^+$ field, the superposition of all individual coil B1+ fields, can be expressed by the law of cosines, a nonlinear function [equation 3, figure 1]. This makes it possible to obtain the value of the B_1^+ of the phase-modulated channel relative to the fixed channel. By weighting this relative value by an absolute B₁⁺ map obtained in the high-flip (HF) angle regime, we obtain the B1+ map of the phasemodulated coil.





the total B_1^+ amplitude $|B_1^+|_{a}$

Method: The relative

B₁ value was extracted from the data using the Fourier series. It is readily seen that decomposition of equation 4 in these series results in the DC and AC component being the a₀ and a₁ Fourier coefficient, respectively. From these coefficients we can calculate the relative B₁⁺. This relative value is weighted using the earlier obtained absolute HF B₁⁺ map steered in quadrature mode, to obtain the absolute B₁⁺ map per channel. This algorithm was tested using a 7T (Achieva, Philips Medical Systems, Best,

the Netherlands) scanner. In vivo measurements were performed using a high-pass birdcage with two ports. In each GRE measurement (FA = 10deg, TR = 50ms, t = 90s) the phase of 1 port was fixed and one was modulated between 0^0 and 360° in 20 steps. For each separate port and for quadrature steering a HF B1+ map (t = 300s) [1] was obtained.

Results: In figure 2A the HF B₁⁺ map in quadrature mode is shown. In 2B and 2D the relative B₁⁺ amplitudes of the two ports are shown, which are based on HF and PMLF measurements respectively. In figure 2C and 2D an FDTD simulated and PMLF measured B₁⁺ map respectively are shown. This demonstrates that the pattern is similar; although an error is seen in the ventricles. This can be caused by long T1 of CSF, which prevents the spins to reach the steady state before the measurement starts.

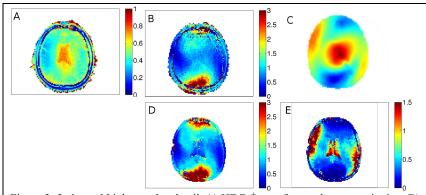


Figure 2: 2 channel high-pass head coil. A) HF B₁⁺ map for quadrature excitation. B) Relative B₁⁺ based on HF B₁⁺ map. C) FDTD simulated B₁⁺ field for one port. D) Relative B₁⁺ based on PMLF GRE images. E) PMLF B₁⁺ map of one port based on 2A and 2D.

- [1] Brunner et al. MRM (2009) 61:1480-1488
- [2] Yarnikh, MRM (2007) 57:192-200
- [3] Van de Moortele, Proc. ISMRM (2009) 367

Discussion and conclusion: We showed a new technique to measure the B1+ field fast and at low SAR cost, which was validated by FDTD simulations and HF B₁⁺ measurements. This method does not require assumptions on the symmetry between B1+ and B1- as is needed in similar low flip angle methods [3]. The shown technique can be expanded to measure the individual B₁ maps of Ncoil arrays. Especially for larger arrays, its speed will be an advantage. A disadvantage of the technique can be the lack of interference between the fixed and modulated channels at distant locations from the phase modulated coil. By working with other coil interference sets (e.g. phase modulating more coils simultaneously), the interference amplitude can be improved, which will improve the B1+ measurement [1].